## Lecture 31

The total transmitted field is thus the sum

$$E_{trans} = \left[1 + r_1 r_2 e^{-i\delta} + (r_1 r_2 e^{-i\delta})^2 + \dots\right] \left[-t_1 t_2 e^{-i\delta_1}\right] E_{inc} = \left[\sum_{l=0}^{\infty} (r_1 r_2 e^{-i\delta})^l\right] \left(-t_1 t_2 e^{-i\delta_1}\right) E_{inc}$$

Now of course  $\ r_1 < 1 \ \ {\rm and} \ r_2 < 1$  , so if we define

$$x = r_1 r_2 e^{-i\delta}$$

Then |x| < 1, so the infinite series converges

$$1 + x + x^{2} + x^{3} + \dots = \sum_{l=0}^{\infty} x^{l} = \frac{1}{1 - x}$$
$$E_{trans} = \frac{-t_{1}t_{2}e^{-i\delta_{i}}}{1 - r_{1}r_{2}e^{-i\delta}}E_{inc}$$

We can get the reflected field in exactly the same way:

$$E_{0r} = r_{1}E_{inc}$$

$$E_{1r} = (it_{1})r_{2}(it_{1})e^{-i\delta}E_{inc}$$

$$E_{2r} = r_{1}r_{2}e^{-i\delta}E_{1r}$$
:
$$E_{Nr} = (r_{1}r_{2}e^{-i\delta})^{N-1}E_{1r}$$

$$E_{refl} = \left[\sum_{l=0}^{\infty} (r_{1}r_{2}e^{-i\delta})^{l}\right] (-t_{1}^{2}r_{2}e^{-i\delta})E_{inc} + r_{1}E_{inc}$$

$$E_{refl} = \left[r_{1} - \frac{-t_{1}^{2}r_{2}e^{-i\delta}}{1 - r_{1}r_{2}e^{-i\delta}}\right]E_{inc}$$

Now lets' consider the transmitted intensity:

$$I_{trans} = \left| \frac{-t_1 t_2 e^{-i\delta_1}}{1 - r_1 r_2 e^{-i\delta}} \right|^2 I_{inc}$$

This is a general expression, but it will be useful to consider more specifically the special case of the symmetrical Fabry-Perot, where  $r_1 = r_2 \equiv r$ , and  $t_1 = t_2 \equiv t = \sqrt{1 - r^2}$ .

$$I_{t r a} = \frac{t^4}{\left|1 - r^2 e^{-i\delta}\right|^2} I$$

Now

$$\begin{aligned} \left| 1 - r^2 e^{-i\delta} \right|^2 &= \left( 1 - R e^{-i\delta} \right) \left( 1 - R e^{i\delta} \right) \\ &= 1 - R \left( e^{-i\delta} + e^{i\delta} \right) + R^2 \\ &= 1 - 2R \cos \delta + R^2 \qquad R = r^2 \\ &= 1 - 2R (1 - 2\sin^2 \frac{\delta}{2}) + R^2 \\ &= \left( 1 - R \right)^2 + 4R \sin^2 \frac{\delta}{2} \end{aligned}$$
$$I_{t r a} = \frac{T^2}{\left( 1 - R \right)^2 \left[ 1 + \frac{4R}{\left( 1 - R \right)^2} + 3r \frac{\delta}{2} \right]} I$$

Def.  $F = \frac{4R}{(1-R)^2}$  = "contrast" (sometimes called the <u>coefficient of finesse</u>)

$$\frac{I_{trans}}{I_{inc}} \frac{T^2}{\left(1-R\right)^2} \left[\frac{1}{1+F \operatorname{sin} \frac{\delta}{2}}\right]_{A \operatorname{iry funct}}$$

Now that when

$$\delta = \frac{2nd\omega}{c} \operatorname{co} \boldsymbol{\varphi}_{t} = 0 \, \pi 2 \, \pi, .4.$$

Then  $\sin^2 \frac{o}{2} = 0$ 

$$\Rightarrow \qquad I_{trans} = \frac{T^2}{\left(1 - R\right)^2} I_{inc} \equiv I_{max}$$

If the system is <u>lossless</u>, so T+R=1 , then  $I_{max} = I_{inc}$ !

$$\Rightarrow \quad \text{Can write} \quad I_{trans} = \left(\frac{1}{1 + F \sin^2 \frac{\delta}{2}}\right) I_{\text{max}}$$

When  $\delta = \pi, 3\pi, 5\pi, \dots, \sin^2 \frac{\delta}{2} = 1 \Longrightarrow$ 

$$I_{t r a} \underset{n}{=} I_{m i} \underset{n}{=} \frac{I_{m a x}}{1+F}$$

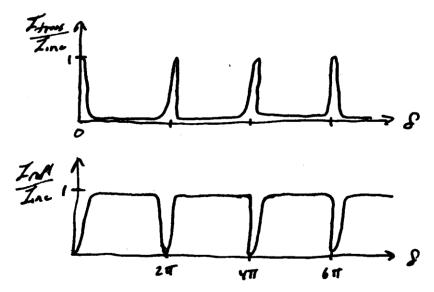
Note that if the mirror reflectance are large, so R is nearly 1, then F becomes very large

$$\Rightarrow I_{\min} \ll I_{\max}$$

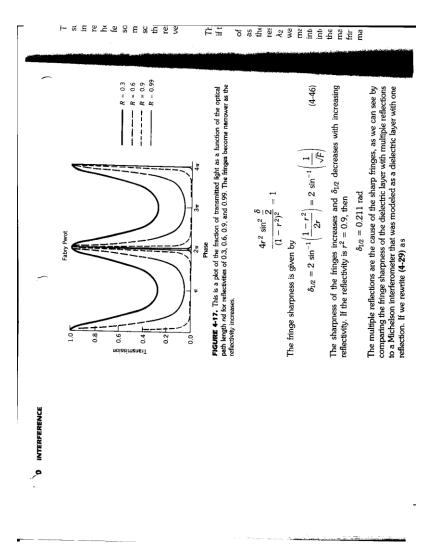
It is easy to verify that for a lossless Fairy-Perot, power is conserved

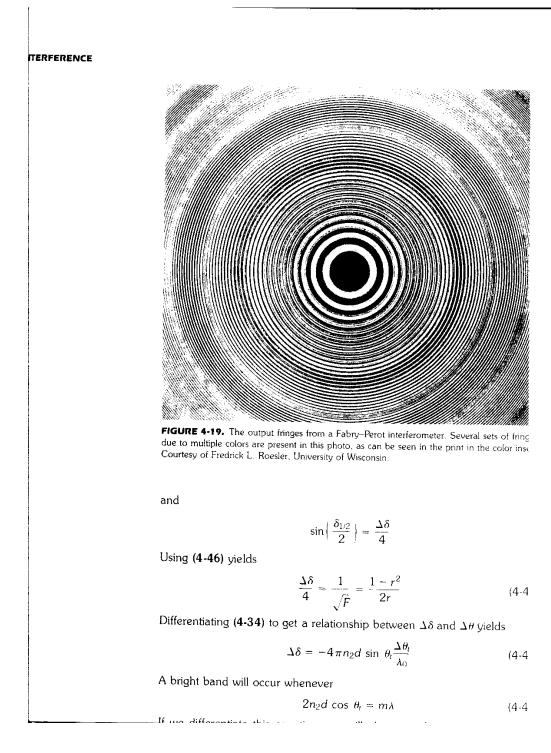
$$I_{inc} = I_{tr} + I_{ans} I,$$

So we have



See Guenther fig.4-17 for accurate plots with different values of R.





Note that the F-P transmission function is periodic, with a maximum every time

$$\frac{2\omega}{c}(n\,d\mathrm{c}\circ\varTheta_{t})=\,\,\mathcal{A}\,\,m\,\,\mathcal{H}\,\,0\,,\,1\,,$$

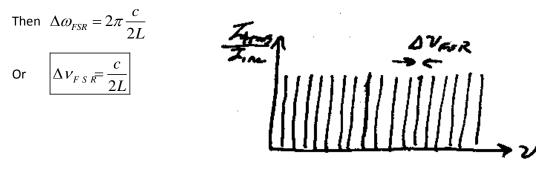
The separation in frequency between two orders  $(\Delta m = 1)$  is called the free spectral range of the Fabry-Perot :

$$\frac{2\Delta\omega_{FSR}}{c} (nd\cos\theta_t) = 2\pi$$

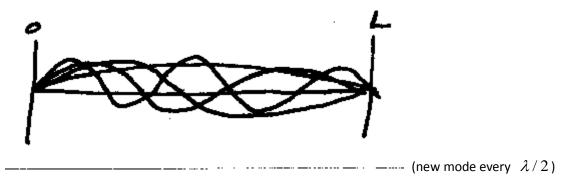
For convenience, we define

 $L = nd \cos \theta_t$ 

This is obviously just the "effective thickness" of the Fabry-Perot! (Not the actual O.P.L.!!)



The free spectral range is often called (especially in laser theory) the <u>axial mode spacing</u> of the Fabry-Perot. This is because the transmission resonances simply correspond to the <u>standing</u> <u>waves</u>, or <u>modes</u>, which the cavity can support.



Now let's consider the <u>width</u> of the transmission resonances. We will characterize the width as usual by the full-width at half-maximum.

i.e. at what values of  $\,\delta\,$  does

$$\frac{1}{1+F\sin^2\frac{\delta}{2}} = \frac{1}{2}$$
We have  $1+F\sin^2\frac{\delta_{1/2}}{2} = 2$ 

$$F\sin^2\frac{\delta_{1/2}}{2} = 1$$

$$\sin\frac{\delta_{1/2}}{2} = \pm\frac{1}{\sqrt{F}}$$
Or  $\delta_{1/2} = \pm 2\sin^{-1}\left(\frac{1}{\sqrt{F}}\right)$ 

Now, when F>>1 (high reflectivity mirrors), as is usually the case , this simplifies considerably , since in this case  $\delta_{v_2}$  is small

$$\sin \frac{\delta_{1/2}}{2} \simeq \frac{\delta_{1/2}}{2} = \pm \frac{1}{\sqrt{F}} \Longrightarrow \delta_{1/2} = \pm \frac{2}{\sqrt{F}}$$
  

$$\Rightarrow \text{ FWHM } \Delta \delta_{1/2} = \frac{4}{\sqrt{F}}$$

It is useful to write this in terms of frequency. Again start with

$$\delta = 2\frac{\omega}{c} (n \, d \, \mathbf{c} \, \mathbf{o} \, \mathbf{e}_{t}) = \frac{\omega}{c} \quad \mathbf{I} = \pi \mathbf{2} \cdot \frac{2L}{c}$$

When F is large,  $\ \Delta\delta$  is small, so are may fake the differential of both sides

$$\Delta \delta = 2\pi \Delta v \cdot \frac{2L}{c}$$

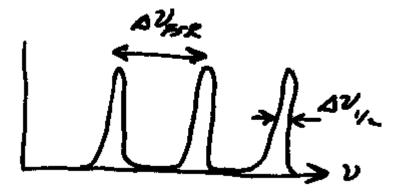
Or 
$$\Delta v_{1/2} = \frac{\Delta \delta_{1/2}}{2\pi} \frac{c}{2L} = \frac{2}{\pi \sqrt{F}} (\Delta v_{FSR})$$

It is traditional to define the finesse  $\mathbb{F}$  by

$$\mathbb{F} = \frac{\pi}{2}\sqrt{F}$$

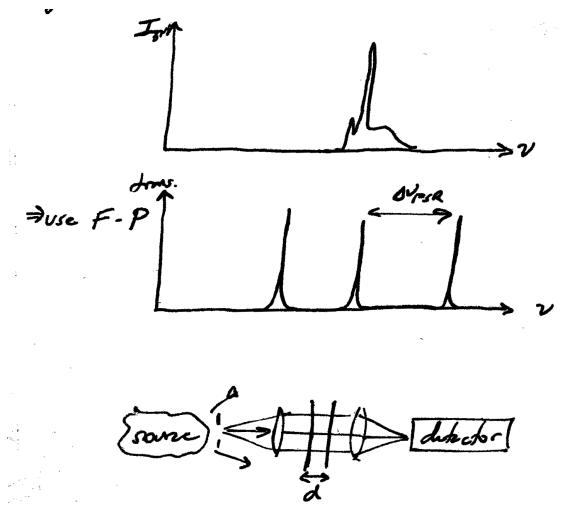
Which yields the ratio of the FWHM to the FSR as

$$\frac{\Delta \delta_{1/2}}{\Delta v_{FSR}} = \mathbb{F}^{-1}$$



Since  $F \gg 1, \mathbb{F} \gg 1$  ,and  $\Delta v_{\rm FSR} \gg \Delta v_{\rm 1/2}$  .

Under these circumstances it is possible to use a Fabry-Perot to perform <u>spectroscopy</u>. e.g. spectrum to be measured



Spectroscopy: measure intensity on detector vs. frequency by scanning the F-P spacing d Note: we want

(i)  $\Delta v_{1/2}$  of the F-P to be ,if possible, <u>narrower</u> than the spectral structure of the source to be

measured

(ii)  $\Delta v_{\rm FSR}$  must be larger than the total extent of the source spectrum (the free spectral range

causes an ambiguity in the frequency otherwise).

A common implementation is to vary the spacing d electrically via piezoelectric crystals separating the mirrors.

Note convolution:  $I_{\text{detector}}(\upsilon) = \int T_{FP}(\upsilon - \upsilon') I_{\text{source}}(\upsilon') d\upsilon'$ 

What is the smallest wavelength or frequency interval that can be measured?

The decision of when two wavelengths are resolved is a qualitative, rather than quantitative one, so the choice has some arbitrariness. One common one is to say that two wavelengths are

resolved when the half-amplitude point of one just overlays the half-amplitude point of the other.



Clearly, the separation is then just  $\Delta 
u_{
m 1/2}$  !

Def. resolving power

$$R = \frac{\lambda}{\Delta \lambda_{1/2}} = \frac{\nu}{\Delta \nu_{1/2}}$$

Note that  $v = m\Delta v_{FSR}$  m=order (mode number)

$$\Rightarrow R = \frac{m\Delta v_{FSR}}{\Delta v_{1/2}} = m\mathbb{F}$$

 $\Rightarrow$  To get a high resolving power, use a high  $\mathbb{F}$  and high m (but you can't increase m forever without running into trouble with the free spectral range!)